Fingerprints of Qubit Noise in Transient Cavity Transmission

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Noise affects the coherence of qubits and thereby places a bound on the performance of quantum computers. We theoretically study a generic two-level system with fluctuating control parameters in a photonic cavity and find that basic features of the noise spectral density are imprinted in the transient transmission through the cavity. We obtain analytical expressions for generic noise and proceed to study the cases of quasistatic, white and $1/f^{\alpha}$ noise in more detail. Additionally, we propose a way of extracting the noise power spectral density in a frequency band only bounded by the range of the qubit-cavity detuning and with an exponentially decaying error due to finite measurement times. Our results suggest that measurements of the time-dependent transmission probability represent a novel way of extracting noise characteristics.

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A main threat to large-scale quantum computation is noise-induced decoherence of the qubit state [1,2]. Common noise sources include fluctuating electromagnetic fields that arise from noisy control parameters as well as the interaction with the environment, e.g., 1/f noise due to two-state fluctuators in the material [3], the nuclear spin bath in semiconductor quantum dots [4–6], and magnetic flux noise, quasiparticles, and two-level fluctuators in superconducting circuits [7–10]. Noise may even affect spin qubits in isotopically purified materials as the spinorbit interaction couples the spin and charge, thus introducing charge noise to the system [11,12]. Relevant examples are hole spins in germanium [13–20].

With quantum error correction still a long way off, the characterization and suppression of noise is of the utmost importance in intermediate-scale quantum devices [21-23]. In this Letter, we propose a novel way of extracting noise characteristics using the cavity transmission which has been used for the dispersive readout of quantum systems [24-27] as well as for the determination of system parameters [28,29] and to detect signatures of the strong coupling regime of cavity quantum electrodynamics [27,30–33]. We go beyond the standard steady-state case and demonstrate that fluctuations affecting the qubit leave a clear trace in the transient transmission. Our model is generally applicable to a wide range of cavity-coupled qubit systems, and in contrast to established noise spectroscopy techniques such as the filter function formalism or relaxometry [34], our method does not require many high-fidelity pulses for dynamical decoupling [35–39], nor do we have to make any assumptions on the noise, e.g., it is not necessary to restrict the analysis to weak or Markovian noise.

We study a single mode of an electromagnetic cavity with frequency ω_c interacting with an infinite number of external modes. A qubit affected by noise is placed inside the cavity and interacts with the cavity mode (Fig. 1). We consider both transverse and longitudinal qubit-photon couplings. In the absence of external driving, in a frame corotating with the probe field at frequency ω_p , and within the rotating wave approximation, only the transverse coupling g is relevant, and the cavity-coupled two-level system is described by the Hamiltonian

$$H = [\omega_q + \delta \omega_q(t) - \omega_p]\sigma_z/2 + \Delta a^{\dagger}a + g(a\sigma_+ + a^{\dagger}\sigma_-),$$
(1)

where $\Delta = \omega_c - \omega_p$ is the cavity-probe detuning, σ_z the Pauli Z matrix, a the photon annihilation operator, and σ_- a ladder operator acting on the qubit. The qubit energy separation is affected by noise, $\omega_q + \delta \omega_q(t)$, and the fluctuating component may be written to leading order as $\delta \omega_q(t) = \lambda \delta X(t)$, where $\delta X(t)$ is the dynamical noise that couples to the qubit control parameter X with strength $\lambda = \partial_X \omega_q |_{\delta X=0}$. The qubit-cavity coupling can also be affected by noise, and we will take this into account further below. As is shown in the Supplemental Material [40], the system is well described by the quantum Langevin equations for the expectation values of the operators σ_- and a,

$$\frac{d\langle\sigma_{-}\rangle}{dt} = -i[\omega_{q} + \delta\omega_{q}(t) - \omega_{p}]\langle\sigma_{-}\rangle - \frac{\gamma}{2}\langle\sigma_{-}\rangle + ig\langle\sigma_{z}\rangle\langle a\rangle,$$
$$\frac{d\langle a\rangle}{dt} = -i\Delta\langle a\rangle - \frac{\kappa}{2}\langle a\rangle - ig\langle\sigma_{-}\rangle + \sqrt{\kappa_{1}}\langle b_{\rm in}(t)\rangle, \qquad (2)$$

where γ is the total noise-independent qubit decoherence rate [41], and $\kappa = \kappa_1 + \kappa_2$ is the total cavity loss rate given by the sum of the rates κ_j at port $j \in \{1, 2\}$. $\langle \sigma_z \rangle$ can depend on time but knowledge of its specific form is not



FIG. 1. A two-level system (qubit, shown in blue) with energy splitting ω_q is affected by noise $\delta\omega_q$ (red) and is placed inside a single-mode electromagnetic cavity with frequency ω_c . Partially transparent mirrors allow for the interaction of the qubit-cavity system with external modes. An input field $b_{in}(t)$ enters at port 1, causing an output field $b_{out}(t)$ that leaves the cavity at port 2, thereby creating a time-dependent transmission through the system from left to right.

required for computing the transmission within the perturbative approach presented below. The bath input field at port 1 is chosen to be a plane wave and hence a constant in the rotating frame, $\langle b_{in}(t) \rangle = \langle b_{in} \rangle$, while we assume no input field to be present at port 2.

By solving the system of coupled differential equations (2), one may obtain the cavity transmission amplitude by employing input-output theory [42–44], $A(t) = \langle b_{out}(t) \rangle / \langle b_{in}(t) \rangle = -\sqrt{\kappa_2} \langle a(t) \rangle / \langle b_{in}(t) \rangle$. An exact solution of Eq. (2) cannot be obtained for generic noise. Working in the regime where the qubit-cavity coupling is small compared to the dominant energy scale, $|g| \ll \eta \equiv$ max{ $|\delta_0|, |\kappa - \gamma|$ } with the unperturbed qubit-cavity detuning $\delta_0 = \omega_c - \omega_q$, we may solve Eq. (2) perturbatively to leading order in g/η ,

$$A(t) = \frac{\sqrt{\kappa_1 \kappa_2}}{i\Delta + \kappa/2} (e^{-i\Delta t - \kappa t/2} - 1) - \frac{\sqrt{\kappa_2}}{\langle b_{\rm in} \rangle} e^{-i\Delta t - \kappa t/2} [\langle a(0) \rangle - ig \langle \sigma_-(0) \rangle \mathcal{I}(t)],$$
$$\mathcal{I}(t) = \int_0^t e^{i\delta_0 t' + (\kappa - \gamma)t'/2 - i\lambda \mathcal{X}(t')} dt'.$$
(3)

Here, $\langle \sigma_{-}(0) \rangle$ and $\langle a(0) \rangle$ are initial conditions, and we introduce the noise integral $\mathcal{I}(t)$ containing the stochastic phase $\mathcal{X}(t) = \int_0^t \delta X(t') dt'$. There are a few remarks in order here: (i) Although we treat *g* perturbatively, this does not mean that our approach does not contain strong coupling cases with $|g| > \kappa$, γ . There, the approximation remains sound if $|g| \ll |\delta_0|$. (ii) To leading order in *g* the long-time solution is unchanged by the noise, and only the transient transmission allows for a determination of noise characteristics. The transient phase generally lasts for a time $\sim 1/\kappa$, but the decay of the noise integral also depends on γ and the details of the noise as discussed below. The minimum

detection time interval T_d decreases with increasing κ_2 and signal-to-noise ratio (SNR), $T_d \sim 1/(\kappa_2 \text{SNR})$ [45,46], and it must be smaller than the typical timescale on which the noise changes, a condition that can be particularly restrictive for high-frequency noise. The number of data points Nattainable in the transient phase is then determined by the SNR of the detector, $N \sim (1/\kappa)/T_d \sim SNR$, and SNRs exceeding 10² have been reported in the solid-state literature [45,47,48]. Additionally, the finite time averaging process dictates a maximum detuning δ_0^m , and for $\kappa \sim \gamma$ we require $g \ll \delta_0^m \sim 1/T_d$ for our results to be valid. (iii) Even at this point we can see the role of the initial qubit state. For $\langle \sigma_{-}(0) \rangle$ to be nonvanishing, we need the qubit to be initialized in a coherent superposition of its energy eigenstates, and in the following we assume $\langle \sigma_{-}(0) \rangle = 1/2$. Moreover, we assume that $\langle a(0) \rangle = 0$, e.g., the cavity may initially be empty. (iv) There are two quantities in Eq. (2) that are affected by finite temperature effects: the qubit level population $\langle \sigma_z(t) \rangle$ and the decoherence rate γ . The former appears in the expansion of $\langle a(t) \rangle$ only at higher orders in perturbation theory, and the latter is only altered in magnitude at increased thermal energies, while the form of the Langevin equations is unchanged. As a result, Eq. (3) also describes the noisy transmission at finite temperature. Remarkably, temperature does not wash out the noise traces in the transient cavity transmission in magnitude. However, an increased γ can lead to a quickly decaying noise signal, hence requiring small measurement times.

Averaging over the noise is possible once we consider an observable quantity, such as the transmission probability $|A|^2$ that will be investigated here. We remark that since the zeroth-order term in Eq. (3) is not affected by noise one has $\langle \langle |A|^2 \rangle \rangle = \langle \langle |A| \rangle \rangle^2$ to first order in g/η , and hence the variance of |A| vanishes. In general, the *k*th central moment of |A| can become nonzero only at order *k* or higher in g/η , implying $\langle \langle |A|^k \rangle = \langle \langle |A| \rangle \rangle^k + \mathcal{O}[(g/\eta)^2]$ [40]. In present-day two-level systems one may expect $\gamma \sim \kappa \sim g \sim MHz$ [27]. For the perturbative approach to be valid we then must consider the dispersive regime, $|\delta_0| \gg |g|$. Assuming symmetric mirrors $\kappa_1 = \kappa_2 = \kappa/2$ and choosing $\langle b_{in} \rangle$ to be real, we obtain up to leading order in g/η ,

$$\langle\!\langle |A(t)| \rangle\!\rangle = |A_{\infty}| \sqrt{\xi_0(t) + \xi_1(t)}.$$
 (4)

Here, $|A_{\infty}| = \kappa/\sqrt{4\Delta^2 + \kappa^2}$ is the Lorentz-shaped transmission through an empty cavity at long times $\kappa t \gg 1$, and we introduce the quantities

$$\xi_{0}(t) = 1 + e^{-\kappa t} - 2e^{-\kappa t/2} \cos \Delta t,$$

$$\xi_{1}(t) = \frac{ge^{-\kappa t/2}}{\sqrt{2/\kappa} \langle b_{in} \rangle} [F(t) \operatorname{Re} \langle\!\langle \mathcal{I}(t) \rangle\!\rangle + G(t) \operatorname{Im} \langle\!\langle \mathcal{I}(t) \rangle\!\rangle],$$

$$F(t) = \frac{2\Delta}{\kappa} (\cos \Delta t - e^{-\kappa t/2}) - \sin \Delta t,$$

$$G(t) = \cos \Delta t - e^{-\kappa t/2} + \frac{2\Delta}{\kappa} \sin \Delta t.$$
(5)

 $\xi_0(t)$ describes the transient signal of an empty cavity, while $\xi_1(t)$ is a correction term due to the interaction with the noisy qubit. The latter can be seen as the normalized fluctuations in the averaged transmission probability,

$$\xi_1 = \frac{\langle\!\langle |A|^2 \rangle\!\rangle - |A_{\infty}|^2 \xi_0}{|A_{\infty}|^2} = \frac{\delta \langle\!\langle |A|^2 \rangle\!\rangle}{|A_{\infty}|^2}, \tag{6}$$

and features the averaged noise integral (ANI),

$$\langle\!\langle \mathcal{I}(t)\rangle\!\rangle = \int_0^t e^{i\delta_0 t' + (\kappa - \gamma)t'/2} \langle\!\langle e^{-i\lambda\mathcal{X}(t')}\rangle\!\rangle dt', \qquad (7)$$

where $\langle \langle \dots \rangle \rangle$ denotes the average over many measurements. Since only first-order processes are taken into account and single photons hence cannot contain noise correlations, only the averaged fluctuations ξ_1 allow for a characterization of noise features. In order to be able to neglect the q^2 term when expanding the absolute value squared of the transmission amplitude (3) while still suppressing higher orders in the perturbation expansion, we require $\sqrt{\kappa}/\langle b_{\rm in}\rangle \sim 1$. This restriction, however, is not severe as $\langle b_{
m in}
angle$ may be tuned externally and independently of the remaining parameters. Equation (4) describes the transmission for quite general systems and without any specifications of the noise $\delta X(t)$ or the corresponding stochastic phase $\mathcal{X}(t)$, and it is shown for exemplary parameter settings in Fig. 2. By recording the noisy part of the transmission via comparison with the transmission through an empty cavity for two distinct detunings Δ_1 and Δ_2 , one may extract the real and imaginary part of the ANI for any δ_0 up to a desired maximum time t_m by choosing $|\Delta_2 - \Delta_2|$ $\Delta_1 | t_m < \pi$ [40].

Up to this point the linearized noise has been treated exactly. In many realistic systems, however, noise affecting ω_q will also affect g. Writing $g \rightarrow g + \delta g(t)$, where to leading order $\delta g(t) = \lambda' \delta X(t)$ with $\lambda' = \partial_X g|_{\delta X=0}$, we may work to first order in δg and integrate by parts to obtain an additional term in the averaged transmission [40],

$$\langle\!\langle |A(t)| \rangle\!\rangle = |A_{\infty}| \sqrt{\xi_0(t) + \xi_1(t) + \xi_1'(t)},$$
 (8)

where $\xi_0(t)$ and $\xi_1(t)$ are as given in Eq. (5) and

$$\xi_1'(t) = \frac{\lambda'}{\lambda} \frac{e^{-\kappa t/2}}{\sqrt{2/\kappa} \langle b_{\rm in} \rangle} \mathcal{F}(t) \tag{9}$$

with

$$\mathcal{F}(t) = e^{(\kappa - \gamma)t/2} \langle\!\langle e^{-i\lambda\mathcal{X}(t)} \rangle\!\rangle [G(t)\cos\delta_0 t - F(t)\sin\delta_0 t] + G(t) \left[\delta_0 \mathrm{Im} \langle\!\langle \mathcal{I}(t) \rangle\!\rangle - \frac{\kappa - \gamma}{2} \mathrm{Re} \langle\!\langle \mathcal{I}(t) \rangle\!\rangle - 1 \right] + F(t) \left[\delta_0 \mathrm{Re} \langle\!\langle \mathcal{I}(t) \rangle\!\rangle + \frac{\kappa - \gamma}{2} \mathrm{Im} \langle\!\langle \mathcal{I}(t) \rangle\!\rangle \right].$$
(10)



FIG. 2. The normalized fluctuations in the averaged transmission probability $\delta \langle \langle |A|^2 \rangle / |A_{\infty}|^2 = \xi_1 + \xi'_1$ as a function of time. We compare the cases of a noise-free (green) and noisy (blue/orange) qubit placed inside the cavity. Solid lines are drawn according to Eqs. (5) (blue) and (9) (orange). Squares are numerical results obtained by averaging over 10³ exact solutions of the full Lindblad equation, allowing for up to 12 cavity photons and assuming normally distributed quasistatic noise with zero mean and standard deviation $\delta X_{\rm rms} = 0.05 \delta_0$. (a) The complete first-order transient curve for the case $g = 0.1\kappa =$ $0.1\gamma = 0.01\delta_0$ at $\Delta = 0$. We find excellent agreement between the analytical and numerical results, even in the presence of a fluctuating coupling constant as can be seen from the magnified section of the plot in panel (b). (c) The initial first-order transient curve for the strong coupling case $g = \kappa = \gamma = 0.01\delta_0$ at $\Delta = 0$. (d) The first-order transient curve for $g = \kappa = 0.01\gamma$ at $\delta_0 = \Delta =$ 0.1 γ with $\delta X_{\rm rms} = \gamma$. The remaining parameter values used are $\langle \sigma_z(0) \rangle = 0, \ \lambda = 0.9, \ \lambda' = -0.1, \ \omega_a/T = 1, \ \text{and} \ \sqrt{\kappa}/\langle b_{\text{in}} \rangle = 1.$

In this case the fluctuations are $\delta \langle \langle |A|^2 \rangle \rangle / |A_{\infty}|^2 = \xi_1 + \xi'_1$ in analogy to Eq. (6). While we must treat the noise in gperturbatively, this is not a severe restriction since the fluctuations in g are expected to be bounded by the coupling itself, $\delta g = \lambda' \delta X < g$. The effect of noise in the energy separation and the qubit photon coupling on the averaged transmission can be clearly seen in Fig. 2(b). At the double resonance $\Delta = \delta_0 = 0$ the term ξ_1 vanishes identically, and for weak coupling ξ'_1 is the dominant contribution.

Having obtained an expression for the measurable average transmission probability for generic longitudinal qubit noise affecting the energy separation and the coupling constant, we proceed to study the averaged phase (AP) $\langle \langle e^{-i\lambda \mathcal{X}(t)} \rangle \rangle$ and the ANI $\langle \langle \mathcal{I}(t) \rangle \rangle$ in more detail. The stochastic phase $\mathcal{X}(t)$ is defined as the time integral over the noise $\delta X(t)$ which is assumed to have zero mean in the remainder of this Letter. When the autocorrelations $\langle \langle \delta X(0) \delta X(\tau) \rangle \rangle$ decay on timescales τ_c which are small compared to the time of integration, the random phase is a sum of many independent random variables. In this situation the central limit theorem guarantees that the probability distribution of $\mathcal{X}(t)$ is Gaussian, and we may write for the AP [49,50]

$$\langle\!\langle e^{-i\lambda\mathcal{X}(t)}\rangle\!\rangle = \exp\left[-\frac{\lambda^2}{2\pi} \int_0^\infty \frac{\sin^2(\omega t/2)}{(\omega/2)^2} S(\omega) d\omega\right], \quad (11)$$

where $S(\omega)$ is the noise spectral density, given by the Fourier transform of the noise autocorrelator. Since the lower integration bound is zero, the Gaussian approximation is not expected to hold at measurement times *t* that are of the same order as the noise correlation time τ_c . Instead, the condition $\tau_c \ll t$ must be met for Eq. (11) to accurately describe the stochastic phase in the transient cavity transmission. On the other hand, $\delta X(t)$ itself may be the sum of many uncorrelated microscopic modes. In this case $\mathcal{X}(t)$ will follow Gaussian statistics regardless of the integration time *t*. If none of the above conditions are met, one must go beyond the Gaussian approximation [51].

There are two prominent special cases for which the ANI in the Gaussian approximation may be explored further analytically, quasistatic noise and white noise. We first consider the case of quasistatic noise [52]. Assuming the total integration time t to be smaller than the time scale on which the quasistatic noise changes, the ANI becomes a Gaussian and may be evaluated,

$$\langle\!\langle \mathcal{I}(t) \rangle\!\rangle_{\rm qs} = \sqrt{\frac{\pi}{2}} \frac{e^{Y^2}}{\lambda \delta X_{\rm rms}} \left[\operatorname{erf}(Y) + \operatorname{erf}\left(\frac{\lambda \delta X_{\rm rms}t}{\sqrt{2}} - Y\right) \right],\tag{12}$$

where $Y = [i\delta_0 + (\kappa - \gamma)/2]/\sqrt{2}\lambda\delta X_{\rm rms}$, $\delta X_{\rm rms} = \sqrt{\langle\!\langle \delta X^2 \rangle\!\rangle}$ is the root mean square of the noise, and erf denotes the error function. We now turn to the case of white noise, where $S = S_0$ is constant. The exponent of the AP becomes linear in time, yielding the exact expression for the ANI,

$$\langle\!\langle \mathcal{I}(t) \rangle\!\rangle_{w} = \frac{e^{i\delta_{0}t + (\kappa - \gamma)t/2 - \lambda^{2}S_{0}t/2} - 1}{i\delta_{0} + (\kappa - \gamma)/2 - \lambda^{2}S_{0}/2}.$$
 (13)

For $S_0 = 0$, Eq. (13) is the ANI in the noise-free case. As white noise describes a Markovian process, it only renormalizes the qubit decoherence rate γ stemming from the Lindblad formalism, $\gamma \rightarrow \gamma + \lambda^2 S_0$.

Next, we investigate generic noise. In real measurements, data cannot be acquired over an infinitely broad frequency band. This can be taken into account by introducing an ultraviolet (uv) cutoff in Eq. (11), i.e., by shifting the upper integration bound from infinity to ω_{uv} . For $\omega_{uv}t \ll 1$, the sine function in Eq. (11) may be expanded around zero. Taking into account the leading term in the expansion, the exponent of the AP becomes quadratic in time and the ANI may be evaluated,

$$\langle\!\langle \mathcal{I}(t) \rangle\!\rangle = \frac{\pi}{\sqrt{2P}} \frac{e^{Z^2}}{\lambda} \left[\operatorname{erf}(Z) + \operatorname{erf}\left(\sqrt{\frac{P}{2\pi}} \lambda t - Z\right) \right], \quad (14)$$

where $Z = \sqrt{\pi/2P} [i\delta_0 + (\kappa - \gamma)/2]/\lambda$ and $P = \int S(\omega)d\omega$ is the noise power in the band $[0, \omega_{uv}]$. As noise in quantum computation must be considered over a large bandwidth corresponding to gate operation times, the noise power *P* provides a practical figure of merit for the comparison of quantum information platforms [53]. Realistically, the condition $\omega_{uv}t \ll 1$ can be fulfilled for spectra dominated by low frequencies such as $1/f^{\alpha}$ noise [54], which is ubiquitous in solid-state systems [3,55,56]. In these cases an additional infrared cutoff ω_{ir} is needed to regularize the power integral [57,58], justified, e.g., by finite data acquisition times [59]. For instance, for $S = C/\omega$ one has $P = C \ln(\omega_{uv}/\omega_{ir})$.

Finally, we consider arbitrary detunings δ_0 . Truncation of the perturbation expansion is valid in the regime $g \ll |\kappa - \gamma|$, which is often realized through $\gamma \gg \kappa \sim g$ in solid-state qubits. Suppose that the ANI has been characterized for at least three values of the noise coupling strength λ which is controllable by external parameters. One may then consider the second derivative of the ANI,

$$\frac{d^2 \langle\!\langle \mathcal{I}(t,\delta_0) \rangle\!\rangle}{d\lambda^2} \Big|_{\lambda=0} = e^{i\delta_0 t + (\kappa - \gamma)t/2} \zeta(t) + \frac{16}{\pi(\kappa - \gamma + 2i\delta_0)} \times \int_0^\infty \frac{S(\omega)}{(\kappa - \gamma + 2i\delta_0)^2 + 4\omega^2} d\omega, \quad (15)$$

where $\zeta(t)$ is a function that is upper bounded by a quadratic scaling in *t*. Hence, the relative error caused by neglecting the first term is upper bounded by the scaling $|\kappa - \gamma|^2 t^2 e^{(\kappa - \gamma)t/2}$, and for measurement times *t* with $\kappa t \sim 1 \ll (\gamma - \kappa)t$ it can be safely neglected, while the effect of the noise on the transmission is still visible [Fig. 2(d)]. By employing a partial fraction decomposition and using the symmetry of $S(\omega) = S(-\omega)$, the nonvanishing part of Eq. (15) may be rewritten as [40]

$$\frac{d^2 \langle\!\langle \mathcal{I}(\delta_0) \rangle\!\rangle}{d\lambda^2}\Big|_{\lambda=0} = \frac{16}{(\kappa - \gamma + 2i\delta_0)^2} \mathcal{C}(\delta_0), \qquad (16)$$

where $C(\delta_0) = (S \star K)(\delta_0)$ denotes the convolution of *S* with the kernel $K(\delta_0) = (\kappa - \gamma + 2i\delta_0)^{-1}$. After Fourier transforming the kernel analytically, we may apply the convolution theorem to obtain

$$S(\omega) = -4 \int_0^\infty \tilde{\mathcal{C}}(\tau) \cos(\omega\tau) e^{(\gamma-\kappa)\tau/2} d\tau, \qquad (17)$$

where $\tilde{C}(\tau)$ denotes the Fourier transform of the measurable convolution. Hence, *S* can be obtained by extracting $C(\delta_0)$ and its Fourier transform from $\langle\!\langle |A| \rangle\!\rangle$ and evaluating (17). A numerically reconstructed spectral density is shown in



FIG. 3. A numerical reconstruction of *S* for quasistatic noise using the parameters in Fig. 2(d) at $\kappa t = 1$. The accuracy is determined by the maximum detuning δ_0^m available in experiment as indicated in the figure. The theoretical result is $S_{qs}(\omega) = 2\pi\delta X_{rms}^2\delta(\omega)$, and the method based on Eq. (17) reproduces this form via the nascent delta function $(\tau_m/\pi) \operatorname{sinc}(\omega \tau_m)$ as the upper integration bound τ_m approaches infinity. The curves are shifted by constant values for clarity.

Fig. 3. Finite detection times do not introduce any additional restrictions on δ_0 and γ , and the results remain valid if $T_d \ll \min\{1/\kappa, 1/\Delta\}$ [40].

Future research may assess the effect of quantum noise on the transmission. In this case *S* can have an antisymmetric contribution and its extraction is not possible using the scheme described in this Letter. Additionally, given the overarching goal of noise mitigation one may investigate whether the qubit coherence in the presence of noise can be protected by the cavity photons.

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